# OPERATING RESULTS OF COMPOUND-TYPE THYRISTOR EXCITATION SYSTEM FOR KANSAI ELECTRIC POWER CO.

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## I. INTRODUCTION

A compound type thyristor static excitation system has been developed for ac synchronous generators, and the first system was delivered to the Takayama Power Station of the Kansai Electric Power Co. Before delivery the system was checked through the various basic tests using the generator in our model transmission equipment. After tests at both the factory and site, the system went into operation in December, 1968.

In order to confirm various problems, characteristic tests including 3-phase short-circuit were carried out in the middle of February this year. The excellent results obtained will be outlined in this report.

# II. OUTLINE OF THE COMPOUND TYPE THYRISTOR EXCITATION SYSTEM

In order to achieve compound characteristics in the thyristor excitation system for synchronous machines, the voltage component related to the generator terminal voltage  $(V_v)$  and the component related to the generator output current  $(V_I)$  are composed vectorically and the resulting vector sum becomes the thyristor input source. In order to control the thyristor firing angle, this power is fed via a current/voltage converter (hereafter referred to as an excitation current transformer) such as a gap CT or a CT+L. Fig. 1 is an outline of this system.

The relation between the voltage component  $\dot{V}_{V}$ 

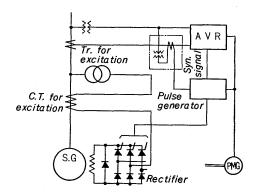


Fig. 1 Thyristor type excitation system with compound performance

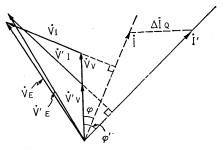


Fig. 2 Vector of thyristor input voltage

and the current component  $V_I$  in Fig. 1, is shown in Fig. 2. The vector sum of  $\dot{V}_V$  (voltage component) and  $\dot{V}_I$  (current component) becomes the thyristor input voltage. However, vector  $\dot{V}_I$  is leading the generator output current vector by 90°. With this vector relation, the thyristor input voltage can be maintained at the required value even if generator terminal voltage is low during bus line shorts or generator overloads. When the generator terminal voltage  $\dot{V}_V$  in Fig. 2 is decreased to  $\dot{V}_V$ , the generator output current increases by  $\Delta \dot{I}_Q$ , the excitation transformer secondary output voltage changes from  $\dot{V}_I$  to  $\dot{V}_I'$ , and  $\dot{V}_E$  becomes  $\dot{V}_E'$ .

If the bus line voltage drops to zero, the vector sum of  $V_I$  due to the corresponding generator output current and the residual voltage  $V_{V'}$  due to the impedance drop in the generator transformer becomes  $V_{E'}$  (at this time, the vectors of  $V_{I'}$  and  $V_{V'}$  are in almost the same phase).

It is possible to achieve compound characteristics for the thyristor input voltage, with the aforementioned construction. The synchronized signal for thyristor firing angle control must naturally be based on  $V_E$  because of the synchronizing between thyristor input voltage and its firing device input voltage. However, the thyristor input voltage waveform is badly distorted due to the impedance of the excitation current and the voltage transformers and the commutation phenomenon in the rectification equipment. For this reason, it is impossible to use thyristor input voltage  $V_E$  directly as a synchronized signal for stable firing angle control. This fact made all previous compound type thyristor excitation system impractical.

This defect was overcome as follows. For the excitation current and voltage transformers in the thyristor input source, small model voltage and current transformers with the same conversion ratios and saturation characteristics were connected to a PT and CT respectively for metering or AVR. The vector sum of the secondary output voltages of these models become the synchronized signal (Refer to Fig. 1).

In this method, the model circuit contains no rectifiers so that there is no distortion in the voltage waveform. Also there is absolutely no trouble of firing angle control since the phases of the voltage vector for the synchronized signal from the model circuit and of the thyristor input basic voltage for main excitation circuit are remained in the same phase under transient conditions during voltage fluctuation as well as during line faults.

#### III. FUNDAMENTAL PRINCIPLES

If the ac side equivalent impedance of the generator field circuit resistor is used as a parameter of the thyristor firing phase  $\alpha$ , the following equation can be applied the excitation circuit shown in Fig. 1.

$$\dot{V}_{V} + \dot{V}_{I} = \dot{I}_{F} Z_{F} + j \dot{I}_{F} (X_{V} + X_{I})$$

$$\dot{\Xi} \dot{V}_{F} + j \dot{I}_{F} \cdot X_{I} \quad \cdots \qquad (1)$$

$$( : V_{F} = \dot{I}_{F} Z_{F}, \quad X_{I} \gg X_{V})$$

where  $V_F$ : Vector of thyristor input voltage or field voltage ac side reduced phase voltage

 $\dot{I}_F$ : Field current ac side reduced phase current

 $Z_F$ : Field reistance ac side reduced impedance per phase, (— coefficient of  $\alpha$  —)

 $V_v$ : Secondary phase voltage of excitation voltage transformer

V<sub>I</sub>: Secondary phase voltage of excitation current transformer

 $X_V$ : Secondary reduced reactance of excitation voltage transformer

 $X_I$ : Secondary reduced reactance of excitation current transformer

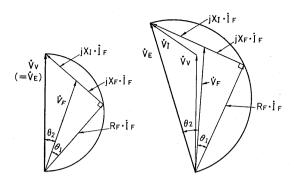
$$\dot{V}_I = j X_I \frac{\dot{I}_1}{b} \qquad (2)$$

where  $\dot{I}_1$ : Generator output current

b: Winding ratio of excitation current transformer

If  $Z_F$  is divided into a resistance component  $R_F$  and a reactance component  $X_F$ , equation 3 results.  $R_F$  and  $X_F$  are both coefficients of the firing angle  $\alpha$ .

The vector relations of equations (1), (2) and (3) are shown in Fig. 3. For the relations between  $R_F$  and  $X_F$  and the firing angle  $\alpha$ , please refer to Vol. 15, No. 3 of the Fuji Electric Review.



(a) No-load condition (b) On-load condition

Fig. 3 Vector diagram of exciting circuit

# IV. FEATURES OF THE COMPOUND TYPE THYRISTOR EXCITATION SYSTEM

This system has the following features

- (1) Compound characteristics can be obtained for the thyristor input voltage. This compound input vector  $\dot{V}_E$  is the vector sum of the voltage component  $\dot{V}_V$  and the secondary output voltage  $\dot{V}_I$  of the current/voltage converter.
- (2) In the previous compound type self-excitation systems, overcompensation was difficult, but in this system it is possible whenever required. Excitation is also maintained even with large voltage drops due to some fault for long periods.

During the boosting operation for generator field from the excitation current transformer, there is current division via the reactor, but if the thyristor firing angle  $\alpha$  is under full gate position  $(\alpha_{\min})$ , the field input is maintained in more than 100% if necessary.

- (3) Field overcurrent limitting equipment is provided in the transistor type AVR, so that optional short-circuit current output of generator can be maintained during 3-phase short-circuits. Furthermore transient response characteristics are ideal and the generator voltage can maintain itself after faults are being eliminated no matter how long the interruption.
- (4) By overcompensation characteristics of the excitation current transformer, an armature winding heating test of generator is possible with the self-excitation circuit only.
- (5) There is no problem during opening of the secondary circuit of excitation current transformer.
- (6) The excitation current transformer is effective in decreasing the amount of waveform distortion in the generator voltage which occurs during thyristor commutation.
- (7) This method is convenient for line charge operation or leading operation of generator, because the firing angle never inclines to  $\alpha_{max}$ .
- (8) Compared with previous compound type selfexcitation system, this system is much more compact.

## V. RESULTS OF MEASUREMENTS

The methods of eliminating the aforementioned problems were confirmed experimentally. An outline will be given here of data obtained from experiments carried out on the acutal equipment at the Takayama

Station of Kansai Electric Power Co. as well as on models. The purpose of these experiments was to back up technical planning work concerning the various features, especially voltage establishment during or after the elimination 3-phase short-circuits.

The specifications for the equipment used in these tests are as follows.

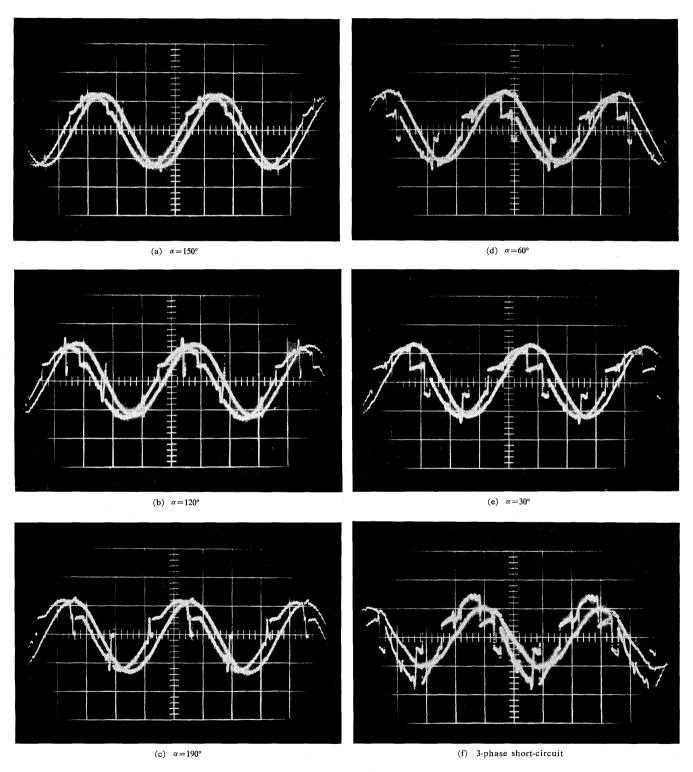


Fig. 4 Waveforms of VF and Vsyn

1) Generator

Rated output: 6500 kva
Rated voltage: 6.6 kv
Frequency: 60 Hz
No. of phases: 3
Rated power factor: 0.9
Speed: 450 rpm

2) Excitation transformer

Primary voltage : 6600 v

Secondary voltage: 181 v (with 290,245,209 and

and 154 v taps) (using 181 v tap)

Capacity : 140 kva No. of phases : 3

(Note: 140 kva is based on the 290 v tap)

3) Exciatation current transformer Primary current: : 569 amp

Secondary voltage: 59 v/phase (with  $59 \times 9/14$ 

and  $59 \times 5/4$  taps) (using the 59 v tap)

Capacity: Primary 57.5 kva, secondary

49.3 kva

No. of phase : 3

4) Rectifier equipment

Connection: 3-phase thyrister-diode mixed bridge

Dc output: 220 v 330 amp

5) AVR

Type: Standard TRANSIDYN PI type AVR with field overcurrent limiting equip-

ment

6) Generator field specifications

No-load field voltage :  $V_{f0} = 95 \text{ v}$ No-load field current :  $I_{f0} = 137 \text{ amp}$ Field voltage during rated output :  $V_{fn} = 181 \text{ v}$ Field current during rated output :  $I_{fn} = 260 \text{ amp}$ Field current during 3-phase short-circuit :

 $I_{fs} = 133 \text{ amp}$ 

Field resistance:  $R_t = 0.65 \Omega$  (75°C)

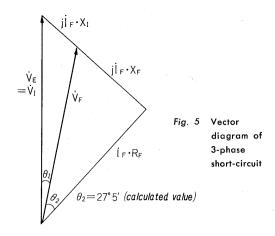
#### 1. Voltage Waveforms in Thyristor Circuit

Fig. 4 shows the voltage waveforms of the thyristor input phase voltage  $\dot{V}_F$  and the synchronized signal voltage  $\dot{V}_{\rm syn}$  at various firing angles, which are obtained in factory tests. Fig. 4 (f) shows the waveforms during a 3-phase short-circuit at the generator terminal. When the output of the excitation current transformer alone serves as the thyristor input souce, the synchronized signal is also obtained only from the output voltage of a model current transformer via a CT for AVR. Phase discrepancies between  $\dot{V}_F$  and  $\dot{V}_{\rm syn}$  are according to phase discrepancies at the measuring points (30°), and it is evident that under all the conditions shown below, the basic wave phase difference between  $\dot{V}_F$  and  $\dot{V}_{\rm syn}$  was almost zero.

Fig. 5 shows the vectors obtained during generator heating operation (factory test) to confirm the vectors in Fig. 3.

Measured values:  $\dot{I}_1 = 729 \text{ amp}$ 

 $\dot{V}_F = 43.8 \text{ v/phase}$  (ac)



$$I_f = 169.2 \text{ amp}$$
 (dc)  
 $V_f = 88 \text{ v}$  (dc)  
 $\dot{V}_F = 41.9 \text{ v/phase}$  (ac)

Excitation voltages transformer: 181 v tap

Excitation current transformer:

569 amp/37.8 v,  $X_I = 0.75 \Omega/\text{phase}$ 

 $(569 \text{ amp}/59 \text{ v.....using } 14 \text{ T tap } 59 \times 9/14 \text{ v}$  using 9 T tap)

The firing angle at this time was  $\alpha_{\min} = 30^{\circ}$ . From the above measured values, the secondary voltage  $\dot{V}_I$  of the excitation current transformer is obtained as follows:

$$\dot{V}_{I} = \frac{37.8}{569} \times 729$$
= 48.4 v .....(4)

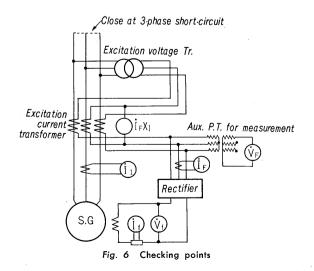
And the field resistance:

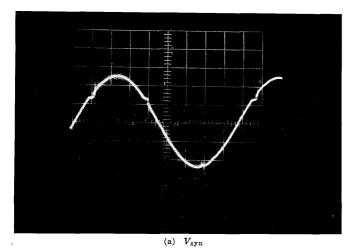
$$R_f = \frac{88}{169.2}$$
  
= 0.52 (\Omega) .....(5)

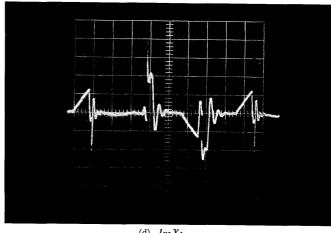
Since  $\alpha = 30^{\circ}$ , the equivalent resistance  $(R_F)$  of  $R_f$  is as follows:

$$R_F = \frac{1}{0.954 (1 + \cos 30)} \times 0.52$$
  
= 0.292 (\Omega) \cdots \tag{6}^{\pm 1}

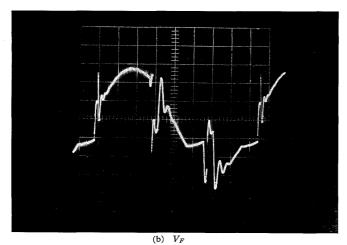
The equivalent power factor of  $Z_F$  at  $\alpha = 30^{\circ}$  is

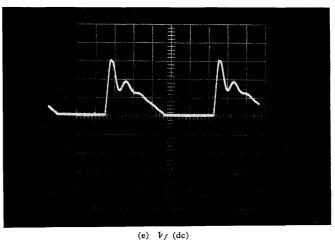






(d)  $I_F \cdot X_I$ 





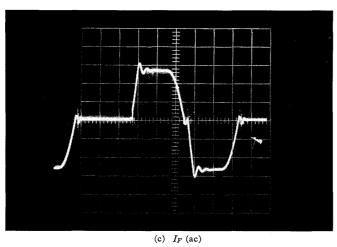


Fig. 7 Waveforms under no-load conditions

$$|\dot{I}_F| = I_f \times \sqrt{\frac{2}{3}}$$
  
= 169.2 × 0.82  
= 139 (amp) ......(9)\*1

\*1 (Note: Refer to Fuji Electric Review Vol. 15, No. 3)

From equations 7, 8 and 9,

$$|\dot{I}_F R_F| = 40.5 \text{ (v)}$$
  
 $|\dot{I}_F X_F| = 20.7 \text{ (v)}$   
 $|\dot{I}_F X_I| = 10.4 \text{ (v)}$  (10)

Fig. 5 was obtained from equation 10. The value of  $|V_I|$  obtained from Fig. 5 coincides with that of equation 4, as well as  $|\dot{V}_F|$  and measured value. Fig. 6 shows the measuring points.

Figs. 7, 8 and 9 show the waveforms at each of the points when the generator is connected in parallel as in Table 1. The generator power output was kept at zero.

given approximately by equation 7.  $\cos \theta = 0.477 (1 + \cos 30)$ 

$$=0.89 \cdots (7)^{*1}$$

From equation 7,  $X_F$  is:  $X_F = R_F \times \tan \theta$ 

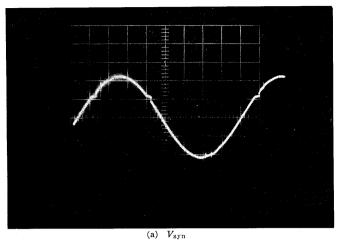
$$=0.292 \times 0.511$$

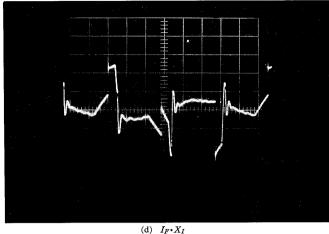
=0.149 
$$(\Omega)$$
 .....  $(8)^{*1}$ 

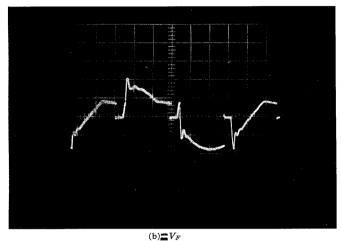
The equivalent effective value  $(|\dot{I}_F|)$  of  $I_f$  can be obtained as a parameter of the firing angle  $\alpha$  as follows.

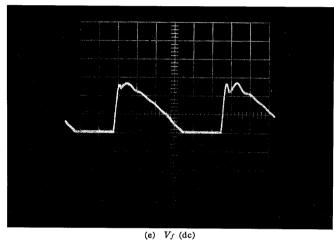
## 2. Synchronizing Operation with Voltage Difference

The synchronizing operation test with some voltage difference between generator terminal voltage and









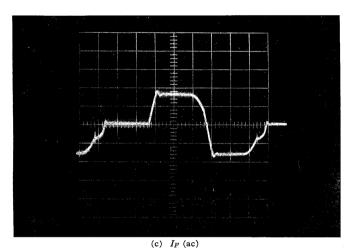


Fig. 8 Waveforms at lagging (2 MVAR lag)

Table 1 Conditions of Waveform Checking under Paralle! Running

Measuring Point Condition	Generator Voltage	Generator Current	Oscillogram
No-Load Parallel	6120	0	Fig. 7
Lagging Reaction Power (2 MVAR)	6450	184	Fig. 8
Leading Reactive Power (1.2 MVAR)	5940	116	Fig. 9

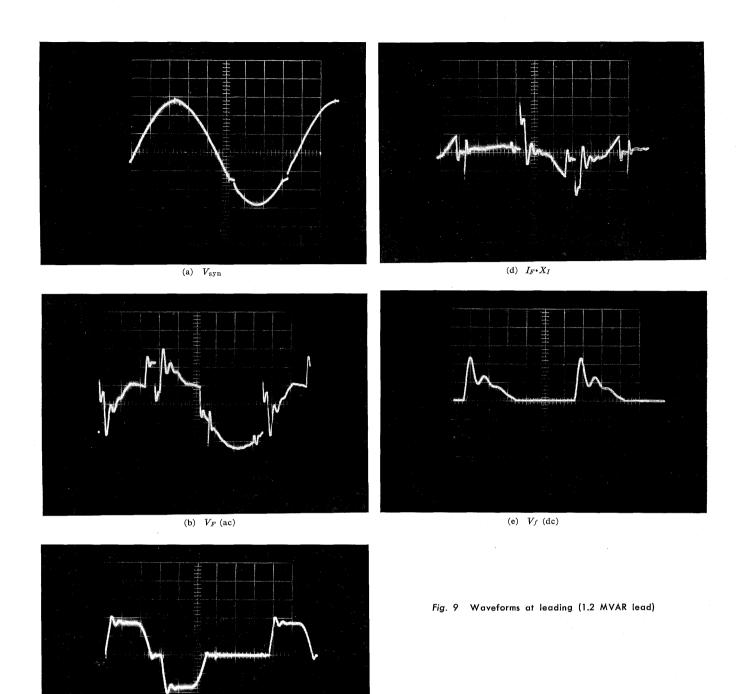
line voltage was done, so as to confirm firing angle control of AVR and to check transient output voltage of the excitation current transformer.

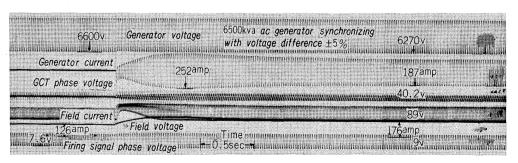
Unlike previous compound type static excitation equipment, voltage source of thyristor is an summation of two voltage vector (v+t). In this condition it was confirmed that the firing angle control at transient condition is never disturbed by any noise, for instance dc component of generator output current at short-circuit fault or the time delay of AVR being caused by a filter of voltage detector circuit.

The test conditions are as shown in *Table 2*, and the oscillograms are shown in *Figs. 10, 11*, and *12*. In *Fig. 12*, the oscillograph speed was increased to show  $V_{\rm syn}$ ,  $V_f$  and  $I_f$  under transient conditions.

# 3. Three-Phase Short-Circuit and the Elimination

One of the main features of compound type thyristor excitation system is that it is possible to establish the voltage itself even after long periods of interruption due to 3-phase short-circuit. However, in order to achieve this requirement, a choice is made for overcompensation of the excitation current transformer on account of the internal voltage drop in the current transformer unit. This means that





(c)  $l_F$  (ac)

Fig. 10 Synchronizing with voltage difference  $V_g > V_L$ 

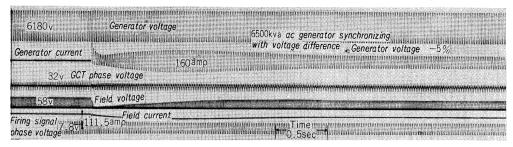


Fig. 11 Synchronizing with voltage difference  $V_g > V^L$ 

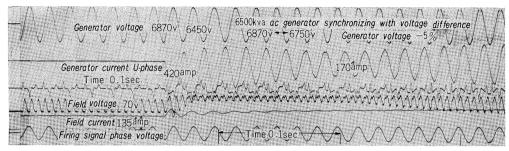


Fig. 12 Synchronizing with voltage difference  $V_{\alpha} > V_{I}$ 

Table 2 Data at Synchronizing with Different Voltage

Condition	Measured Value	V <sub>L</sub> * (v)	$V_g$ * (v)	i <sub>1</sub> (amp)	<i>Ďv</i> (v)人	$\dot{I}_F X_I$ (v) $\downarrow$	$V_F$ (v) $igwedge$	$_{-}I_{F}$ ac	$I_f$ dc
+5% (10 cm/s)	Before	104.5	110.0	0	92.0	34.0	82.0	92.0	126.0
	After	107.5	108.0	187	90.0	40.2	98.5	132	176.0
-5% (10 cm/s)	Before	108.5	103.0	0	94.0	32.0	86.9	80.0	111.5
	After	107.0	105.2	88	92.0	31.0	79.0	60.0	91.3
+5% (100 cm/s)	Before	109.0	114.5	0	94.0	36.0	84.0	89.5	134.5
	After	111.8	112.5	208	98.0	43.0*	103.0	144.0	191.0

<sup>\*) 6600/110</sup> v PT secondary voltage

the short-circuit current during a 3-phase short-circuit is at least maintained at a value above that of the rated generator output current. For overcompensation, the thyristor output current curve "m", in relation to the generator output current obtained only by the excitation current transformer at  $\alpha_{\min}$  must be below the 3-phase short-circuit curve "n" as shown in Fig. 13. However, the generator output current and field current during a 3-phase short-circuit increase until the two curves m and n intersect at the point P as shown in Fig. 13. But it is very difficult to determine the point P beforehand, because it differs greatly according to the saturation characteristics of the excitation current transformer and the generator field resistance. To solve this problem, the AVR has field overcurrent suppression properties ACR, and any field limiting current can be controlled according to set value. In this manner, the generator output current can be suppressed indirectly. Fig. 14 shows  $I_{lmax}$  and  $I_{fACR}$  as the maximum current values for the generator and field respectively. For these

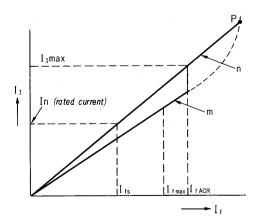


Fig. 13 ACR characteristics

ACR characteristics, it is not necessary to use special manufacturing conditions concerning the saturation characteristics of the excitation current/voltage converter. The set value  $I_{fACR}$  for suppression of field current must be set slightly more than  $I_{fmax}$  which

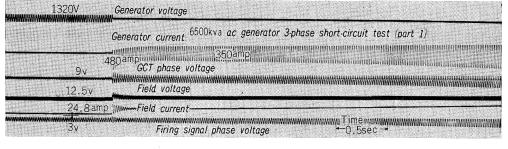


Fig. 14 Oscillogram of 3-phase short-circuit

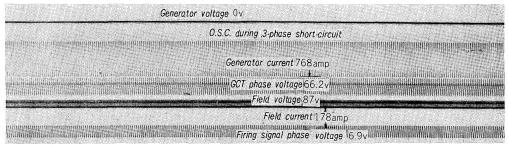


Fig. 15 Oscillogram
under shortcircuit conditions

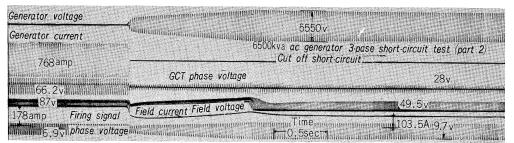


Fig. 16 Oscillogram
at cut-out of

Table 3 Data of 3-phase Short-Circuit Tests

Measurement Point Time of Measurement	$V_g = V_g = V_g$	$I_g$ (amp)	$I_F ullet X_I \ (\mathrm{v}) igwedge$	, $V_F$ (v)人	$I_F$ ac (amp)	$I_f$ (amp)
Before Closing	22.0		9	16.2	8	24.8
After Closing Setting	_	768	66.2	51.0	156	178
After Elimination Setting	89.0		28	65.0	67.8	104

<sup>\* 6600/110</sup> v PT secondary voltage

is required during emergencies. During a normal condition, the ACR has no relation to field exciting control. After the field current increases because of an AVR signal during bus line voltage drop, this value reaches the set  $I_{fACR}$  value, after the characteristic PI response time, the operation mode is switched over from the AVR to the ACR. This switching is not dependent on any external changeover signal, which means that very smooth switching between the AVR and ACR is possible. Table 3 shows the data obtained from 3-phase short-circuit tests. Figs. 14, 15 and 16 show oscillograms at the time of the 3-phase short-circuit, during the shortcircuit and at the time the short-circuit is cut out respectively. The pre-set value of the generator voltage for the short-circuit tests was 20% before the shortcircuit, and imediately after short-circuit the setting voltage is changed over to 80% before cutting out of the short-circuit.  $I_{fACR}$  was 115% of  $I_{fs}$ .

#### VI. CONCLUSION

This article has given an outline of the results of tests carried out in the factory and at the site on the compound type thyristor excitation system which was introduced in the Fuji Electric Review Vol. 15, No. 3. This excitation equipment can maintain all its functions even during line faults and is more economical than other systems. We expect that this system will be used widely as main excitation equipment to improve the power supply reliability and quality in ever-expanding power systems.

In this article excitation system is explained using thyristor-diode mixed bridge construction for the thyristor bridge. If full bridge thyristors are used, by inverter operation, energy of field current will be fed back to ac network through excitation current transformer.

#### Reference

 Kaneda, Moriyasu, Yokokawa: Compound Type Thyristor Excitation System for Large Capacity Generators, Fuji Electric Review, Vol. 15, No. 3